

## DEPARTMENT OF ELECTRONICS AND COMMUNICATION ENGINEERING

### VII Semester

### AU3008 Sensors and Actuators

#### UNIT – I - INTRODUCTION TO MEASUREMENTS AND SENSORS

#### 1.8 Mathematical model of transducers

#### - Zero, First and Second order transducers

#### Mathematical Model of Transducers:

- The mathematical models are the differential equations that describe the dynamics of transducers.
  - ❖ These models can be derived from the knowledge of the components, their interconnection and the physical laws governing their functioning.
  - ❖ A number of assumptions are needed to derive the equations representing the model.
  - ❖ But practically, the components used, their values, their behaviour, their interconnections and the physical laws followed by them may not be precisely known.
  - ❖ Therefore, using conventional method, the model cannot be obtained.
  - ❖ In such situations, the transducer can be assumed to be a black box, whose inputs and outputs are accessible for measurements.
  - ❖ Number of methods are available to identify the transducer model by measuring the inputs and outputs of the transducer.
  - ❖ If the order of the model is known already, then the method of identification becomes simple.

**(1) Identification of transducer mathematical models:**

**Identification from Impulse response**

- Let the transfer function of the transducer be of the form  $\frac{K}{(1 + \tau s)}$  where  $K$  and  $\tau$  are the only unknown parameters.
- When this transducer is excited with an input impulse, the output transform

$$Y(s) = \frac{K}{(1 + \tau s)}$$

as

$$R(s) = 1$$

- Therefore  $y(t) = L^{-1} Y(s)$

$$= Ke^{-t/\tau} \quad \dots (2.45)$$

- The output of the transducer is shown in fig. (2.20)

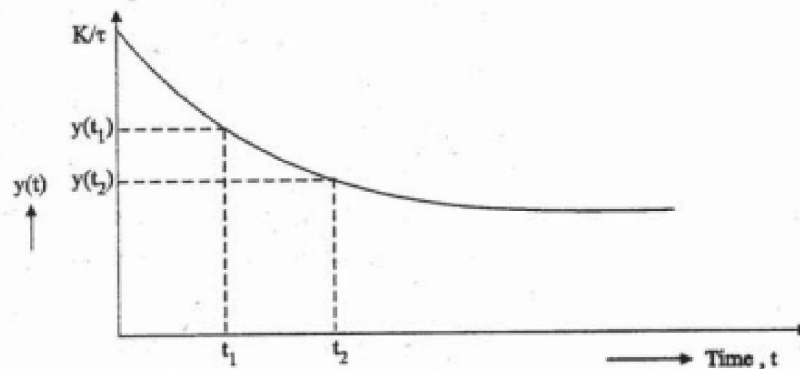


Fig. 2.20 First-order transducer response for impulse signal.

- From the experimentally obtained outputs of the transducer  $y(t_1)$  and  $y(t_2)$  at two different times  $t_1$  and  $t_2$ , the two unknown parameters  $K$  and  $\tau$  of the transducer can be estimated.

$$y(t_1) = \frac{K}{\tau} e^{-t_1/\tau} \quad \dots (2.46)$$

$$y(t_2) = \frac{K}{\tau} e^{-t_2/\tau} \quad \dots (2.47)$$

$$\frac{y(t_1)}{y(t_2)} = e^{(t_2 - t_1)/\tau} \quad \dots (2.48)$$

$$T = \frac{(t_2 - t_1)}{\ln \frac{y(t_1)}{y(t_2)}}$$

- $K$  can be calculated by substituting  $\tau$  in one of the above equations.
- If the transfer function is of the form

$$\frac{K}{\frac{s^2}{\omega_n^2} + \frac{2\xi}{\omega_n}s + 1}$$

- where  $\xi$ ,  $\omega_n$  and  $K$  are the unknown parameters.
- When such a system is subjected to an unit impulse, the response for the underdamped case will be as shown in fig. (2.21)

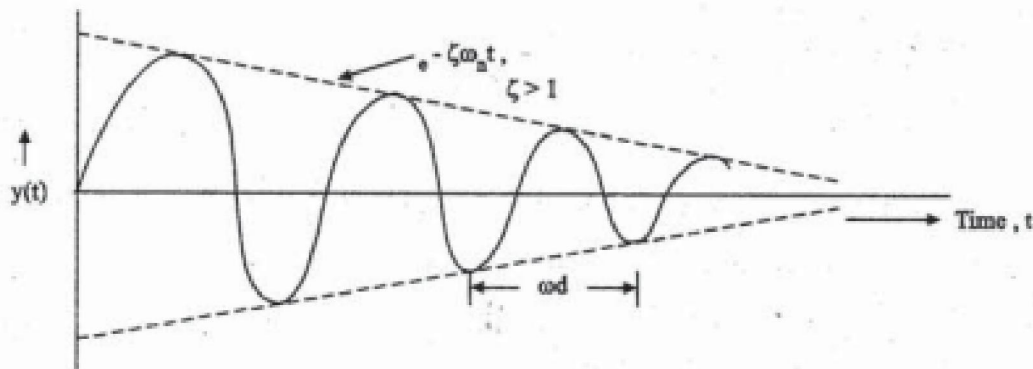


Fig. 2.21. Response of II - order transducer for impulse input signal.

$$y(t) = \frac{K\omega_n}{\sqrt{1-\xi^2}} e^{-\xi\omega_n t} \sin \omega_n [\sqrt{1-\xi^2} t] \quad \dots (2.49)$$

- From the experimental output curve,  $\xi$ ,  $\omega_n$  is calculated taking the envelope (dotted line) only.
- As the envelope is a decaying exponential curve,  $\frac{1}{\xi\omega_n}$  is the time constant of the exponential curve.
- The time between two successive peaks  $T_d$  is determined which is equal to

$$\omega_d = \frac{2\pi}{T_d} = \omega_n \sqrt{1-\xi^2} \quad \dots (2.50)$$

- Now the values of  $\xi$ , and  $\omega_n$  are determined from the values of  $\frac{1}{\xi\omega_n}$  and the value of the envelope at  $t = 0$  is

$$\frac{K\omega_n}{\sqrt{1 - \xi^2}} \quad \dots (2.51)$$

which can be determined from the experimental response.

- As  $\xi$  and  $\omega_n$  have already been evaluated,  $K$  can be calculated.

## (2) Identification from step response

- When the transfer function of the transducer is of the form  $\frac{K}{(1 + \tau s)}$ , the parameters  $K$  and  $\tau$  have to be determined from the step response.
- The static sensitivity  $K$  is calculated as

$$K = \frac{\text{Steady state output change}}{\text{Input change}} \quad \dots (2.52)$$

- For a second-order transducer, the parameters,  $K$ ,  $\xi$ , and  $\omega_n$  can be determined from the step response.
- The response of an under damped transducer for an unit step input is shown in fig. (2.22).
- The expression for the output  $y(t)$  of the transducer is given by

$$y(t) = K \left[ 1 - \frac{e^{-\xi\omega_n t}}{\sqrt{1 - \xi^2}} \right] \sin(\omega_n \sqrt{1 - \xi^2} t + \phi) \quad \dots (2.53)$$

- The time instances at which the maximum and minimum values of the response curve occur can be found out by differentiating  $y(t)$  with respect to time and equating to zero as shown below.

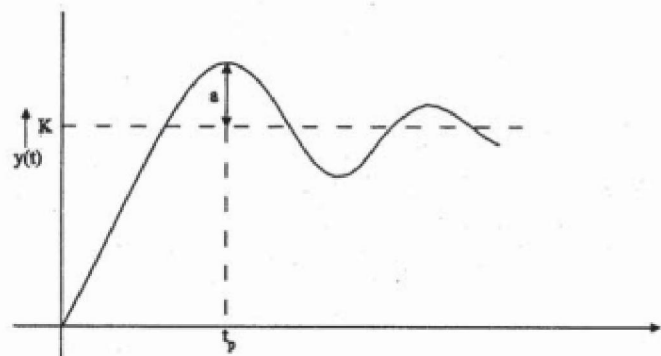


Fig. (2.22) second - order transducer response for step input.

$$\frac{dy(t)}{dt} = K \left\{ \xi \omega_n \frac{e^{-\xi \omega_n t}}{\sqrt{1-\xi^2}} \sin(\omega_n \sqrt{1-\xi^2} t + \phi) - \frac{e^{-\xi \omega_n t}}{\sqrt{1-\xi^2}} \cdot \cos(\omega_n \sqrt{1-\xi^2} t + \phi) \cdot \omega_n \sqrt{1-\xi^2} \right\} \quad \dots (2.54)$$

When this expression is equated to zero, one gets,

$$\tan(\omega_n \sqrt{1-\xi^2} t + \phi) = \frac{\sqrt{1-\xi^2}}{\xi}$$

- $\phi = \tan^{-1} \frac{\sqrt{1-\xi^2}}{\xi^2}$  (by definition)

- Therefore,  $\tan(\omega_n \sqrt{1-\xi^2} t + \phi) = \tan \phi$  ... (2.55)

- This equation is true for all values of

$$t = 0, \frac{\pi}{\omega_n \sqrt{1-\xi^2}}, \frac{2\pi}{\omega_n \sqrt{1-\xi^2}}, \frac{3\pi}{\omega_n \sqrt{1-\xi^2}} \text{ etc.} \quad \dots (2.56)$$

- when  $t = 0$ ,  $y(t)$  is 0, minimum value

$$t = t_p = \frac{\pi}{\omega_n \sqrt{1-\xi^2}} \quad \dots (2.57)$$

- $y(t)$  is the first maximum and  $t_p$  is the peak time.

$$t = t_v = \frac{2\pi}{\omega_n \sqrt{1-\xi^2}} \quad \dots (2.58)$$

- $y(t)$  - second minimum and  $t_v$  - valley time.

- As the oscillation is a damped one, the time at which the first maximum occurs will be the maximum overshoot.

- Therefore this overshoot shown as ' $\alpha$ ' in fig. (2.22) can be obtained as

$$\alpha = y(t)|_{\max} - y(t)|_{\text{steady state}}$$

- $y(t)_{\max}$  is obtained by substituting

$$t_p = \frac{\pi}{\omega_n \sqrt{1-\xi^2}} \text{ in equation (2.59)}$$

i.e.,

$$y(t)|_{\max} = K \left\{ \frac{1 - e^{-\xi\omega_n \frac{\pi}{\omega_n \sqrt{1-\xi^2}}}}{\sqrt{1-\xi^2}} \sin \left( \omega_n \sqrt{1-\xi^2} \cdot \frac{\pi}{\omega_n \sqrt{1-\xi^2}} + \phi \right) \right\} \dots (2.60)$$

$$= K \left\{ 1 + \frac{e^{-\pi\xi/\sqrt{1-\xi^2}}}{\sqrt{1-\xi^2}} \sin \phi \right\} \dots (2.61)$$

$$= K \left[ 1 + e^{-\pi\xi/\sqrt{1-\xi^2}} \right] \text{ as } \sin \phi = \sqrt{1-\xi^2} \dots (2.62)$$

$$\begin{aligned} y(t)|_{\text{steady state}} &= \lim_{t \rightarrow \infty} y(t) \\ &= K(1-0) \\ &= K \end{aligned}$$

$$\therefore \text{Overshoot, } \alpha = K e^{-\pi\xi/\sqrt{1-\xi^2}} \dots (2.63)$$

- From the step response plotted from experimental results,  $t_p$ ,  $\alpha$  and  $K$  can be obtained from equation (2.52).  $\xi$  can be calculated from equation (2.63) as  $\alpha$  and  $K$  are already known.
- Substituting this value of  $\xi$  in equ. (2.58),  $\omega_n$  can be determined.

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