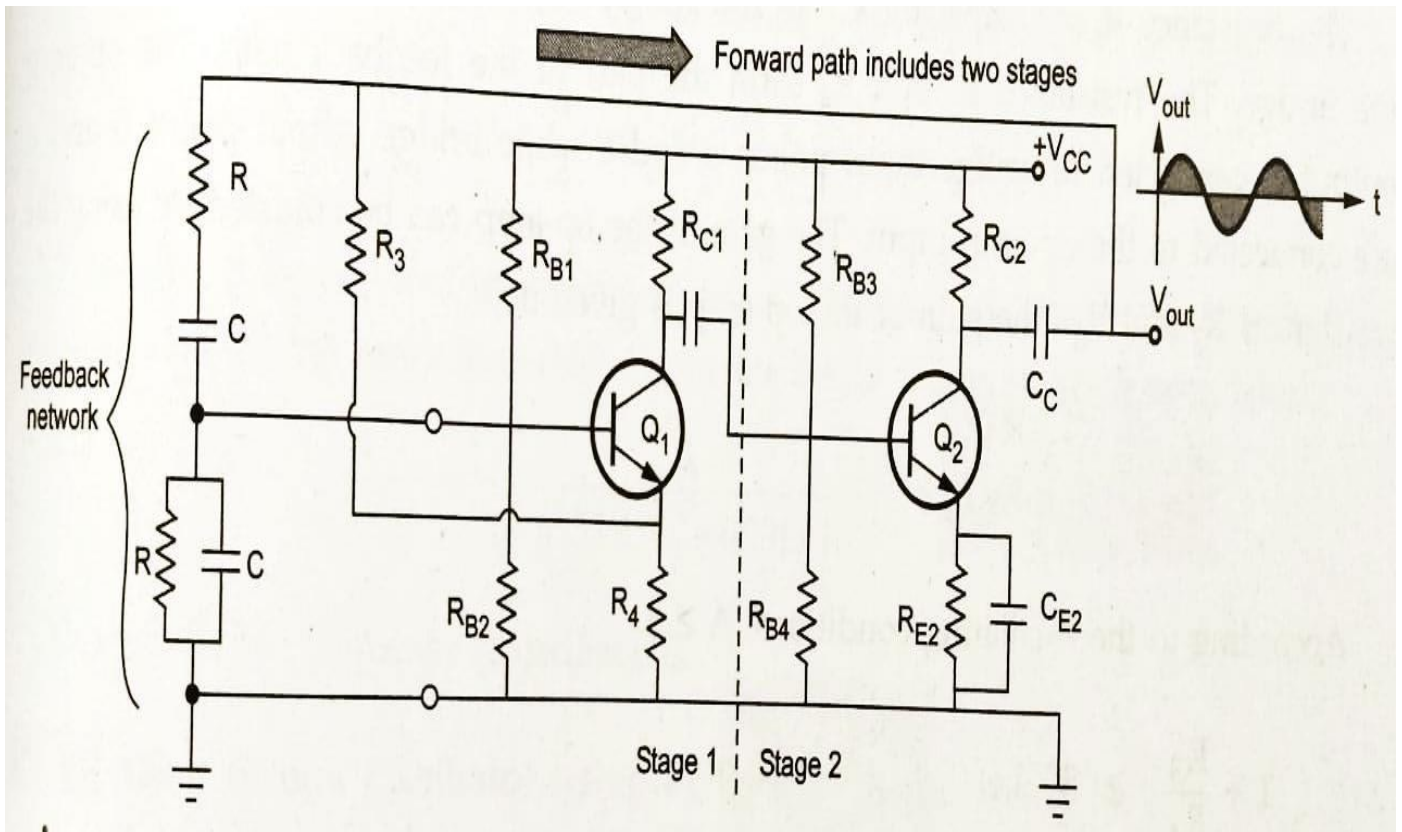


## WIEN BRIDGE OSCILLATOR

- It uses a non-inverting amplifier (does not provide any phase shift during amplifier stage).
- As total phase shift required is  $0^\circ$  or  $2n\pi$  radians, in wien bridge type no phase shift is necessary through feedback.
- Thus the total phase shift around a loop is  $0^\circ$ .
- A Wien-Bridge Oscillator is a type of phase-shift oscillator which is based upon a Wien-Bridge network comprising of four arms connected in a bridge fashion.
- Here two arms are purely resistive while the other two arms are a combination of resistors and capacitors.
- In particular, one arm has resistor and capacitor connected in series ( $R_1$  and  $C_1$ ) while the other has them in parallel ( $R_2$  and  $C_2$ ).
- Two arms of the bridge  $R_1$ ,  $C_1$  in series and  $R_2$ ,  $C_2$  in parallel are frequency sensitive.
- In this circuit, at high frequencies, the reactance of the capacitors  $C_1$  and  $C_2$  will be much less due to which the voltage  $V_0$  will become zero as  $R_2$  will be shorted.
- At low frequencies, the reactance of the capacitors  $C_1$  and  $C_2$  will become very high. However even in this case, the output voltage  $V_0$  will remain at zero only, as the capacitor  $C_1$  would be acting as an open circuit.
- This kind of behavior exhibited by the Wien-Bridge network makes it a lead-lag circuit in the case of low and high frequencies, respectively

Transistorised wien bridge oscillator:

- In this circuit two stage common emitter transistor amplifiers is used.
- Each stage contributes  $180^\circ$  phase shift hence the total phase shift due to the amplifier stage becomes  $360^\circ$  which is necessary as per the oscillator conditions.
- The bridge consists of R and C in series, R and C in parallel,  $R_3$  and  $R_4$ .
- The feedback is applied from the collector of  $Q_2$  through the coupling capacitor, to the bridge circuit.
- The two stage amplifier provides a gain much more than 3 and it is necessary to reduce it.

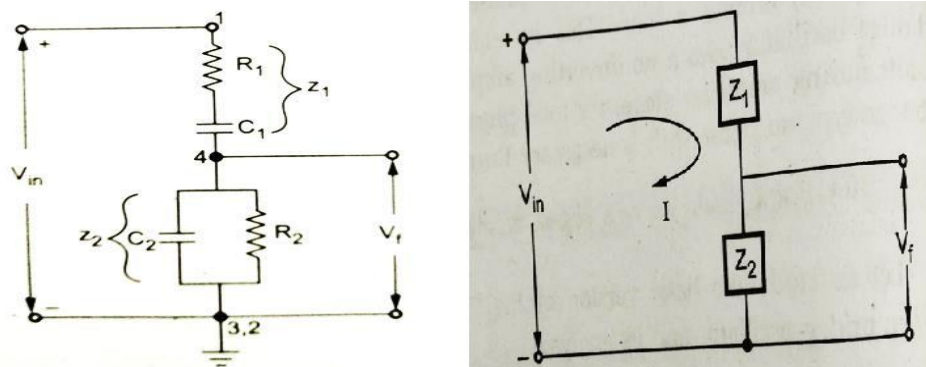


**Figure.1 wien bridge oscillator**

[Source: Microelectronics by J. Millman and A. Grabel,]

- To reduce the gain, the negative feedback is used without bypassing the resistance  $R_4$ .
- The amplitude stability can be improved using a nonlinear resistor for  $R_4$ .
- Increase in the amplitude of the oscillations, increases the current through nonlinear resistance, which results into an increase in the value of non linear resistance  $R_4$ .

Derivation of wien bridge oscillator



**Figure.2 feedback network of wien bridge oscillator**

[Source: Microelectronics by J. Millman and A. Grabel,]

- From figure.2

$$Z_1 = R_1 + \frac{1}{j\omega C_1} = \frac{1 + j\omega R_1 C_1}{j\omega C_1}$$

$$Z_2 = R_2 \parallel \frac{1}{j\omega C_2} = \frac{R_2}{1 + j\omega R_2 C_2}$$

Replaing  $j\omega = s$ ,

$$Z_1 = \frac{1 + sR_1C_1}{sC_1}$$

$$Z_2 = \frac{R_2}{1 + sR_2C_2}$$

$$I = \frac{V_{in}}{Z_1 + Z_2}$$

And  $V_f = IZ_2$

$$V_f = \frac{V_{in}Z_2}{Z_1 + Z_2}$$

$$\beta = \frac{V_f}{V_{in}} = \frac{Z_2}{Z_1 + Z_2}$$

- Substituting the value of  $Z_1$  and  $Z_2$

$$\beta = \frac{\frac{R_2}{1+sR_2C_2}}{\frac{1+sR_1C_1}{sC_1} + \frac{R_2}{1+sR_2C_2}}$$

- Replacing  $s$  by  $j\omega$ ,  $s^2 = -\omega^2$  and rationalizing simplifying the expression

$$\beta = \frac{\omega^2 C_1 R_2 (R_1 C_1 + R_2 C_2 + C_1 R_2) + j\omega C_1 R_2 (1 - \omega^2 R_1 R_2 C_1 C_2)}{(1 - \omega^2 R_1 R_2 C_1 C_2)^2 + \omega^2 (R_1 C_1 + R_2 C_2 + C_1 R_2)^2}$$

- To have zero phase shift of the feedback network, its imaginary part must be zero

$$\omega C_1 R_2 (1 - \omega^2 R_1 R_2 C_1 C_2) = 0$$

$$\omega^2 = \frac{1}{R_1 R_2 C_1 C_2}$$

$$\omega = \frac{1}{\sqrt{R_1 R_2 C_1 C_2}}$$

$$f = \frac{1}{2\pi \sqrt{R_1 R_2 C_1 C_2}}$$

- This is the frequency of the oscillator and it shows that the components of the frequency sensitive arms are the deciding factors for the frequency.
- In practice  $R_1=R_2=R$  and  $C_1=C_2=C$

$$f = \frac{1}{2\pi RC} ; \omega = \frac{1}{RC}$$

$$\beta = \frac{1}{3}$$

- The positive sign of  $\beta$  indicates that the phase shift by the feedback network is  $0^\circ$

$$|A\beta| \geq 1$$

$$|A| \geq \frac{1}{|\beta|} \geq \frac{1}{\frac{1}{3}}$$

$$|A| \geq 3$$

- This the required gain of the amplifier stage without any phase shift.
- If  $R_1 \neq R_2$  and  $C_1 \neq C_2$  then

$$f = \frac{1}{2\pi\sqrt{R_1R_2C_1C_2}}$$

$$\beta = \frac{C_1R_2}{R_1C_1 + R_2C_2 + C_1R_2}$$

$$A \geq \frac{R_1C_1 + R_2C_2 + C_1R_2}{C_1R_2}$$

Advantages:

1. Mounting the two capacitors on common shaft and varying their values, the frequency can be varied as per the requirement.
2. The stability high
3. The frequency range can be selected simply by using decade resistance box
4. High gain